

6.003: Signals and Systems**Fourier Transform**

April 6, 2010

Mid-term Examination #2

Tomorrow, April 7, 7:30-9:30pm.

No recitations tomorrow.

Coverage: Lectures 1–15
 Recitations 1–15
 Homeworks 1–8

Homework 8 will not be collected or graded. Solutions are posted.

Closed book: 2 pages of notes ($8\frac{1}{2} \times 11$ inches; front and back).

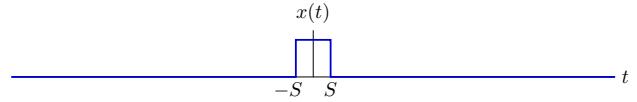
Designed as 1-hour exam; two hours to complete.

Last Week: Fourier SeriesRepresenting periodic signals as sums of **sinusoids**.→ new representations for systems as **filters**.

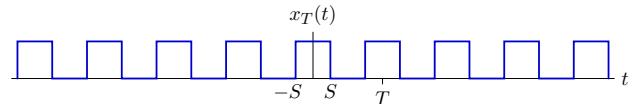
This week: generalize for aperiodic signals.

Fourier Transform

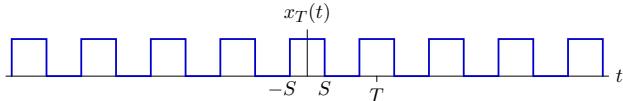
An aperiodic signal can be thought of as periodic with infinite period.

Let $x(t)$ represent an aperiodic signal.

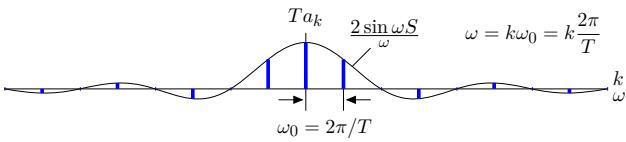
$$\text{"Periodic extension": } x_T(t) = \sum_{k=-\infty}^{\infty} x(t + kT)$$



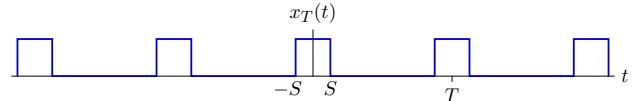
$$\text{Then } x(t) = \lim_{T \rightarrow \infty} x_T(t).$$

Fourier TransformRepresent $x_T(t)$ by its Fourier series.

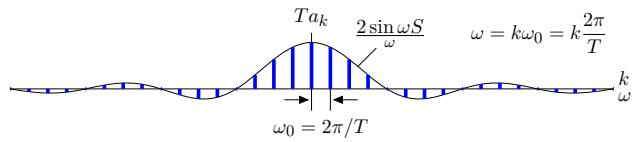
$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x_T(t) e^{-j \frac{2\pi}{T} kt} dt = \frac{1}{T} \int_{-S}^S e^{-j \frac{2\pi}{T} kt} dt = \frac{\sin \frac{2\pi k S}{T}}{\pi k} = \frac{2 \sin \omega S}{T} \frac{1}{\omega}$$

**Fourier Transform**

Doubling period doubles # of harmonics in given frequency interval.

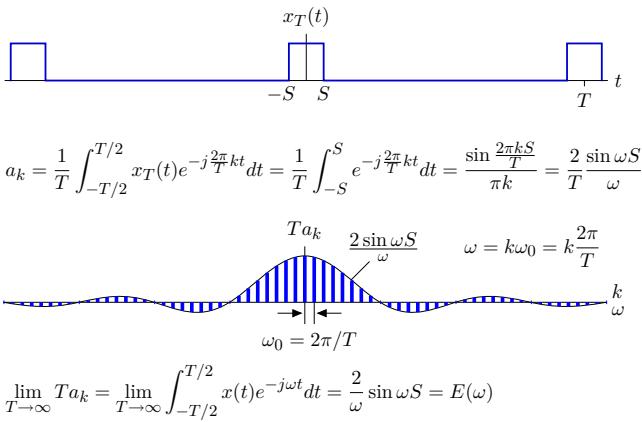


$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x_T(t) e^{-j \frac{2\pi}{T} kt} dt = \frac{1}{T} \int_{-S}^S e^{-j \frac{2\pi}{T} kt} dt = \frac{\sin \frac{2\pi k S}{T}}{\pi k} = \frac{2 \sin \omega S}{T} \frac{1}{\omega}$$

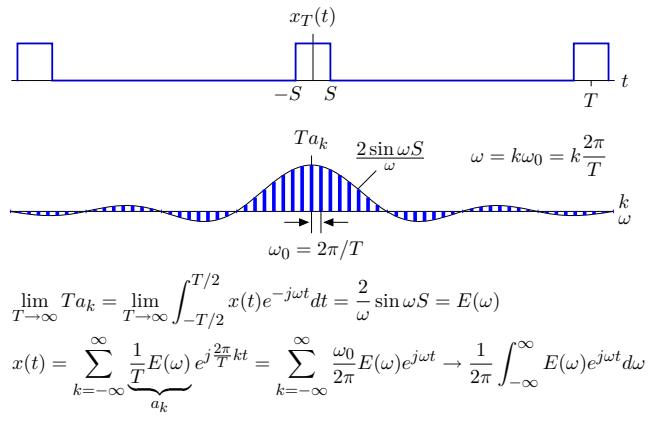


Fourier Transform

As $T \rightarrow \infty$, discrete harmonic amplitudes \rightarrow a continuum $E(\omega)$.

**Fourier Transform**

As $T \rightarrow \infty$, synthesis sum \rightarrow integral.

**Fourier Transform**

Replacing $E(\omega)$ by $X(j\omega)$ yields the Fourier transform relations.

$$E(\omega) = X(s)|_{s=j\omega} \equiv X(j\omega)$$

Fourier transform

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad (\text{"analysis" equation})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \quad (\text{"synthesis" equation})$$

Form is similar to that of Fourier series
 \rightarrow provides alternate view of signal.

Relation between Fourier and Laplace Transforms

If the Laplace transform of a signal exists and if the ROC includes the $j\omega$ axis, then the Fourier transform is equal to the Laplace transform evaluated on the $j\omega$ axis.

Laplace transform:

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

Fourier transform:

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = H(s)|_{s=j\omega}$$

Relation between Fourier and Laplace Transforms

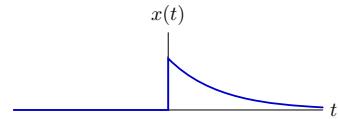
Fourier transform "inherits" properties of Laplace transform.

Property	$x(t)$	$X(s)$	$X(j\omega)$
Linearity	$ax_1(t) + bx_2(t)$	$aX_1(s) + bX_2(s)$	$aX_1(j\omega) + bX_2(j\omega)$
Time shift	$x(t - t_0)$	$e^{-st_0} X(s)$	$e^{-j\omega t_0} X(j\omega)$
Time scale	$x(at)$	$\frac{1}{ a } X\left(\frac{s}{a}\right)$	$\frac{1}{ a } X\left(\frac{j\omega}{a}\right)$
Differentiation	$\frac{dx(t)}{dt}$	$sX(s)$	$j\omega X(j\omega)$
Multiply by t	$tx(t)$	$-\frac{d}{ds} X(s)$	$-\frac{1}{j} \frac{d}{d\omega} X(j\omega)$
Convolution	$x_1(t) * x_2(t)$	$X_1(s) \times X_2(s)$	$X_1(j\omega) \times X_2(j\omega)$

Relation between Fourier and Laplace Transforms

There are also important differences.

Compare Fourier and Laplace transforms of $x(t) = e^{-t} u(t)$.



Laplace transform

$$X(s) = \int_{-\infty}^{\infty} e^{-t} u(t) e^{-st} dt = \int_0^{\infty} e^{-(s+1)t} dt = \frac{1}{1+s}; \text{ Re}(s) > -1$$

a complex-valued function of **complex** domain.

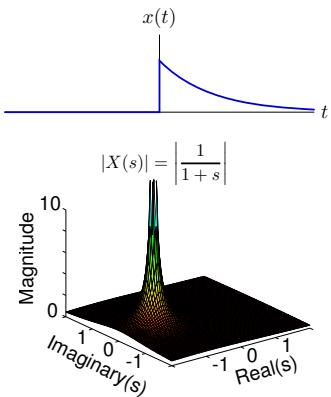
Fourier transform

$$X(j\omega) = \int_{-\infty}^{\infty} e^{-t} u(t) e^{-j\omega t} dt = \int_0^{\infty} e^{-(j\omega+1)t} dt = \frac{1}{1+j\omega}$$

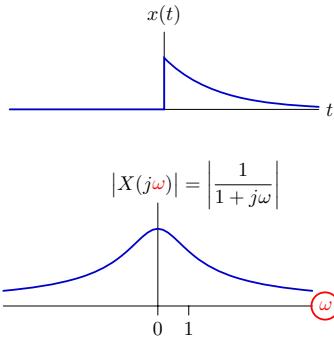
a complex-valued function of **real** domain.

Laplace Transform

The Laplace transform maps a function of time t to a complex-valued function of complex-valued domain s .

**Fourier Transform**

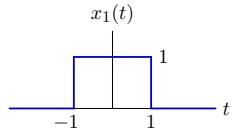
The Fourier transform maps a function of time t to a complex-valued function of real-valued domain ω .



Frequency plots provide intuition that is difficult to otherwise obtain.

Check Yourself

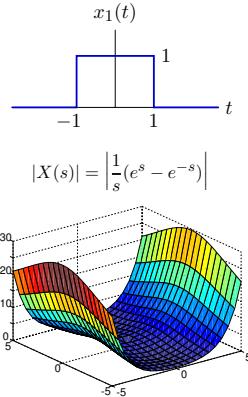
Find the Fourier transform of the following square pulse.



1. $X_1(j\omega) = \frac{1}{\omega} (e^\omega - e^{-\omega})$
2. $X_1(j\omega) = \frac{1}{\omega} \sin \omega$
3. $X_1(j\omega) = \frac{2}{\omega} (e^\omega - e^{-\omega})$
4. $X_1(j\omega) = \frac{2}{\omega} \sin \omega$
5. none of the above

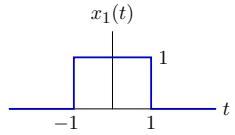
Laplace Transform

Laplace transform: complex-valued function of complex domain.

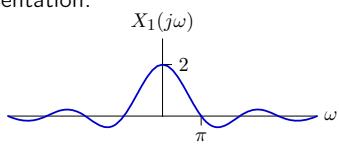
**Fourier Transform**

The Fourier transform is a function of real domain: frequency ω .

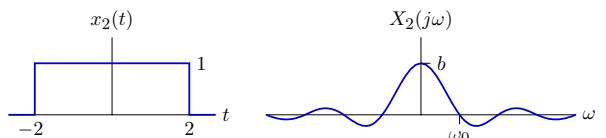
Time representation:



Frequency representation:

**Check Yourself**

Signal $x_2(t)$ and its Fourier transform $X_2(j\omega)$ are shown below.



Which is true?

1. $b = 2$ and $\omega_0 = \pi/2$
2. $b = 2$ and $\omega_0 = 2\pi$
3. $b = 4$ and $\omega_0 = \pi/2$
4. $b = 4$ and $\omega_0 = 2\pi$
5. none of the above

Check Yourself

Stretching time compresses frequency.

Find a general scaling rule.

Let $x_2(t) = x_1(at)$.

If time is stretched in going from x_1 to x_2 , is $a > 1$ or $a < 1$?

Fourier Transforms

Find a general scaling rule.

Let $x_2(t) = x_1(at)$.

$$X_2(j\omega) = \int_{-\infty}^{\infty} x_2(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} x_1(at) e^{-j\omega t} dt$$

Let $\tau = at$ ($a > 0$).

$$X_2(j\omega) = \int_{-\infty}^{\infty} x_1(\tau) e^{-j\omega\tau/a} \frac{1}{a} d\tau = \frac{1}{a} X_1\left(\frac{j\omega}{a}\right)$$

If $a < 0$ the sign of $d\tau$ would change along with the limits of integration. In general,

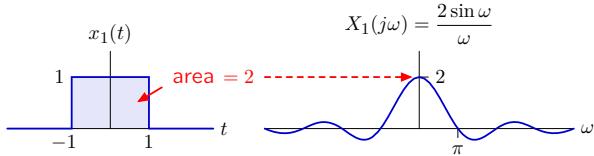
$$x_1(at) \leftrightarrow \frac{1}{|a|} X_1\left(\frac{j\omega}{a}\right).$$

If time is stretched ($a < 1$) then frequency is compressed and amplitude increases (preserving area).

Moments

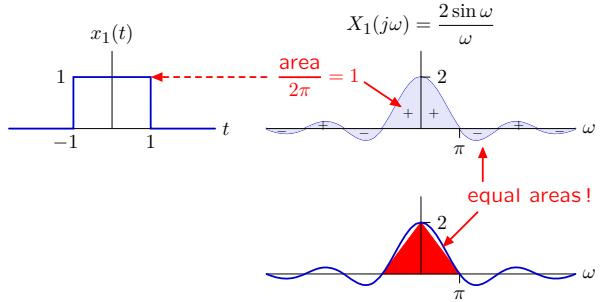
The value of $X(j\omega)$ at $\omega = 0$ is the integral of $x(t)$ over time t .

$$X(j\omega)|_{\omega=0} = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} x(t) e^{j0t} dt = \int_{-\infty}^{\infty} x(t) dt$$

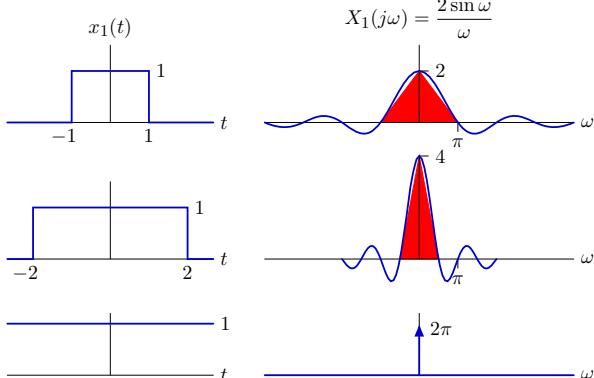
**Moments**

The value of $x(0)$ is the integral of $X(j\omega)$ divided by 2π .

$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega 0} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) d\omega$$

**Stretching to the Limit**

Stretching time compresses frequency and increases amplitude (preserving area).

**Fourier Transform**

One of the most useful features of the Fourier transform (and Fourier series) is the simple “inverse” Fourier transform.

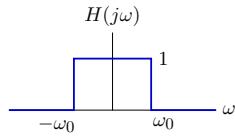
$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad (\text{Fourier transform})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \quad (\text{“inverse” Fourier transform})$$

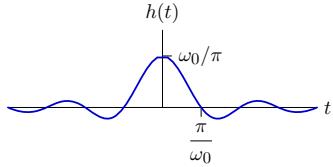
New way to think about an impulse!

Inverse Fourier Transform

Find the impulse response of an “ideal” low pass filter.



$$h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{\omega_0} e^{j\omega t} d\omega = \frac{1}{2\pi} \left. \frac{e^{j\omega t}}{jt} \right|_{-\omega_0}^{\omega_0} = \frac{\sin \omega_0 t}{\pi t}$$



This result is not so easily obtained without inverse relation.

Fourier Transform

The Fourier transform and its inverse have very similar forms.

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad (\text{Fourier transform})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \quad (\text{“inverse” Fourier transform})$$

Convert one to the other by

- $t \rightarrow \omega$
- $\omega \rightarrow -t$
- scale by 2π

Duality

The Fourier transform and its inverse have very similar forms.

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

Two differences:

- minus sign: flips time axis (or equivalently, frequency axis)
- divide by 2π (or multiply in the other direction)

$$x_1(t) = f(t) \leftrightarrow X_1(j\omega) = g(\omega)$$

~~$\omega \rightarrow t$~~ ~~$t \rightarrow \omega$~~ ; flip ; $\times 2\pi$

$$x_2(t) = g(t) \leftrightarrow X_2(j\omega) = 2\pi f(-\omega)$$

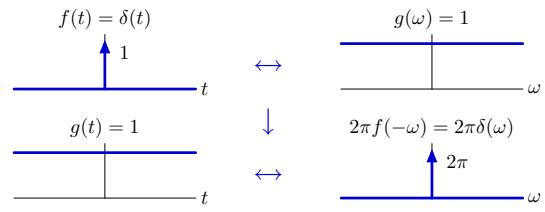
Duality

Using duality to find new transform pairs.

$$x_1(t) = f(t) \leftrightarrow X_1(j\omega) = g(\omega)$$

~~$\omega \rightarrow t$~~ ~~$t \rightarrow \omega$~~ ; flip ; $\times 2\pi$

$$x_2(t) = g(t) \leftrightarrow X_2(j\omega) = 2\pi f(-\omega)$$



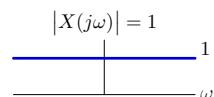
The function $g(t) = 1$ does not have a Laplace transform!

More Impulses

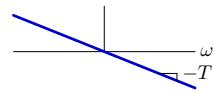
Fourier transform of delayed impulse: $\delta(t-T) \leftrightarrow e^{-j\omega T}$.

$$x(t) = \delta(t-T)$$

$$X(j\omega) = \int_{-\infty}^{\infty} \delta(t-T) e^{-j\omega t} dt = e^{-j\omega T}$$



$$\angle X(j\omega) = -\omega T$$

**Eternal Sinusoids**

Using duality to find the Fourier transform of an eternal sinusoid.

$$\delta(t-T) \leftrightarrow e^{-j\omega T}$$

~~$\omega \rightarrow t$~~ ~~$t \rightarrow \omega$~~ ; flip ; $\times 2\pi$

$$e^{-jtT} \leftrightarrow 2\pi\delta(\omega+T)$$

$T \rightarrow \omega_0 :$

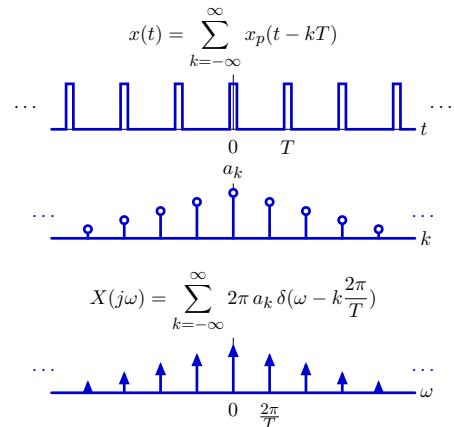
$$e^{-j\omega_0 t} \leftrightarrow 2\pi\delta(\omega+\omega_0)$$

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j \frac{2\pi}{T} kt} \quad \text{CTFS} \quad \longleftrightarrow \quad \{a_k\}$$

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j \frac{2\pi}{T} kt} \quad \text{CTFT} \quad \longleftrightarrow \quad \sum_{k=-\infty}^{\infty} 2\pi a_k \delta\left(\omega - \frac{2\pi}{T} k\right)$$

Relation between Fourier Transform and Fourier Series

Each term in the Fourier series is replaced by an impulse.

**Mid-term Examination #2**

Tomorrow, April 7, 7:30-9:30pm.

No recitations tomorrow.

Coverage: Lectures 1–15
Recitations 1–15
Homeworks 1–8

Homework 8 will not be collected or graded. Solutions are posted.

Closed book: 2 pages of notes ($8\frac{1}{2} \times 11$ inches; front and back).

Designed as 1-hour exam; two hours to complete.

MIT OpenCourseWare
<http://ocw.mit.edu>

6.003 Signals and Systems

Spring 2010

For information about citing these materials or our Terms of Use, visit: <http://ocw.mit.edu/terms>.